

Discrete Fourier transform (DFT) representation assumes that any arbitrary, finite-time signal $y[n]$ may be represented as a sum of sinusoidal signals,

$$y[n] = \sum_{k=0}^{N-1} A_k \cos(\omega_k n + \theta_k), \quad n = 0, \dots, L-1$$

N - ממד התווים

$k = 0, \dots, N-1$

$$\omega_k = k \frac{2\pi}{N}$$

פונקציית $y[n]$ שוגג $N \geq L$ גוראה: $\hat{y} = \hat{y}$

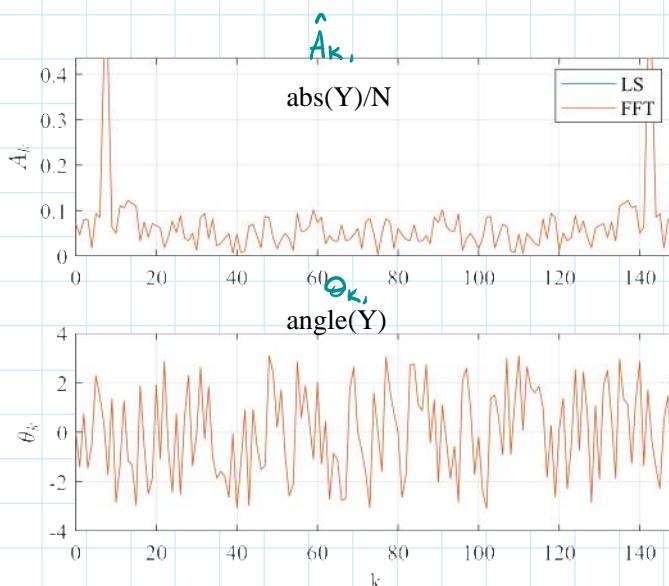
$$\underline{\omega} = \begin{bmatrix} \omega_0 \\ \omega_1 \\ \omega_2 \\ \vdots \\ \omega_{N-1} \end{bmatrix} \rightarrow A_0 \quad \theta_0 = 0 \leq DC$$

$X =$

$$\begin{bmatrix} 1 & \cos(\omega_0 \cdot 0) & \sin(\omega_0 \cdot 0) & \cos(\omega_2 \cdot 0) & \dots \\ 1 & \cos(\omega_1 \cdot 0) & \sin(\omega_1 \cdot 0) & \cos(\omega_3 \cdot 0) & \dots \\ 1 & \cos(\omega_2 \cdot 0) & \sin(\omega_2 \cdot 0) & \cos(\omega_4 \cdot 0) & \dots \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 1 & \cos(2\omega_0) & \sin(2\omega_0) & \cos(2\omega_2) & \dots \end{bmatrix} \quad \text{מגדיר } L \times 2N-1$$

DFT \rightarrow גוראה ישר מהר
במאות סכום

$$Y[k] = \sum A_k e^{j\theta_k}$$



w0 = 0.1*pi;
A = 1.5;
theta = -pi/4;
L = 100;
n = (0:L-1);
sigma = 1;

y_theory = A*cos(w0*n+theta);
y = y_theory + sigma*randn(L,1);

N = L+50; % number of frequencies
%% LS
X = zeros(L,2*N-1);
X(:,1) = ones(L,1);
X(:,2:end) = cos(2*pi/N*n*(1:N-1));
X(:,3:2:end) = sin(2*pi/N*n*(1:N-1));
w_ls = lsqminnorm(X,y); % w_ls = pinv(X)*y;
y_hat = X*w_ls;

%% Amplitude and phase

A_hat(1) = w_ls(1);
A_hat(2:N) = sqrt(w_ls(2:2:end).^2 + w_ls(3:2:end).^2);
theta_hat(1) = 0; DC
theta_hat(2:N) = -atan2(w_ls(3:2:end), w_ls(2:2:end));
Y = fft(y,N);

Frequency estimation

$$y[n] = A \cos(\omega_0 n + \theta) + \epsilon[n] \quad n = 0, \dots, L-1$$

$$\hat{y} = \underline{X} \underline{w}$$

$\hat{y}(\omega_0) = \underline{X}(\omega_0) \underline{w} \leftarrow \text{periodogram}$

$$\mathbf{e}(\omega_0) = \mathbf{y} - \hat{\mathbf{y}}(\omega_0)$$

$$\mathbf{e} \perp \hat{\mathbf{y}}$$

$$\langle \mathbf{y} \rangle^2 = \langle \hat{\mathbf{y}}(\omega_0) \rangle^2 + \langle \mathbf{e}(\omega_0) \rangle^2$$

טוקן פוליאו

טוקן גראן
 ω_0

טוקן אוניברסיטאי

$$\hat{\omega}_0 = \arg \min_{\omega_0} \|\mathbf{e}\|^2 = \arg \max_{\omega_0} \|\hat{\mathbf{y}}\|^2$$

$$\widehat{SNR} = \frac{\|\hat{\mathbf{y}}\|^2}{\|\mathbf{e}\|^2}$$

טוקן SNR
0.828

$$\hat{\sigma}_\epsilon^2 = \|\hat{\mathbf{e}}\|^2 \cdot \frac{1}{L}$$

טוקן ε²
0.828
טוקן ε²
פער 8

$$\langle x^2(t) \rangle = \sigma_x^2$$

dataset
 $w_0 = 0.1 * \pi;$
 $A = 1.5;$
 $\theta = -\pi/4;$
 $L = 100;$
 $n = (0:L-1)';$
 $\sigma = 1;$
 $y_theory = A * \cos(w_0 * n + \theta);$
 $noise = \sigma * randn(L, 1);$
 $y = y_theory + noise;$
 $snr_theory = y_theory' * y_theory / (noise' * noise);$

LS
function $P = per(y, w_0)$
 $L = length(y);$
 $n = (0:L-1)';$
 $X = [\cos(w_0 * n) \sin(w_0 * n)];$
 $w_ls = lsqminnorm(X, y);$
 $y_hat = X * w_ls;$
 $P = y_hat' * y_hat;$
end
fun = @(w) -per(y, w);
w_max = fminsearch(fun, 0.1 * pi);

$$P_{\cos} = \frac{A^2}{2}$$

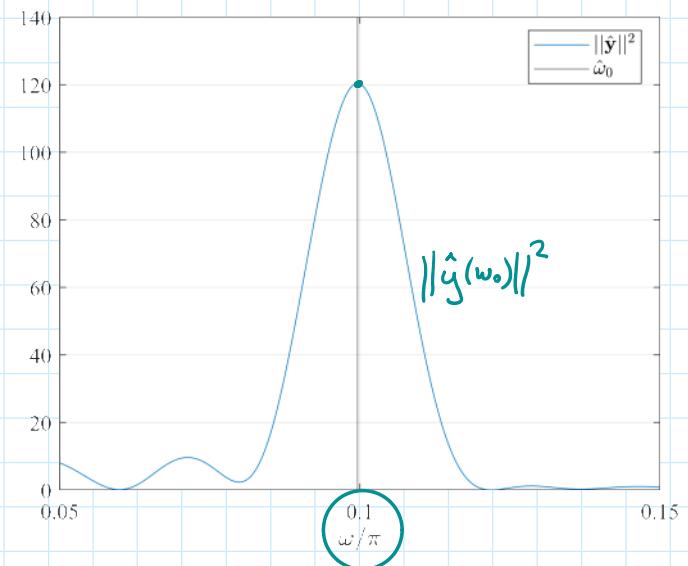
טוקן אוניברסיטאי
 $\hat{A} - A$

$$SNR = \frac{A^2}{2\sigma^2} \quad (19.25)$$

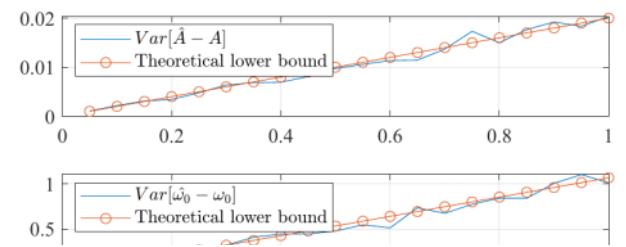
and the corresponding (Cramer-Rao) bounds are

$$\text{Var}[\hat{A}] \geq \frac{2\sigma^2}{L} [V^2] \quad (19.26)$$

$$\text{Var}[\hat{\omega}_0] \geq \frac{12}{SNR \times L(L^2 - 1)} \approx \frac{12}{SNR \times L^3} \left[\left(\frac{\text{rad}}{\text{sample}} \right) \right] \quad (19.27)$$

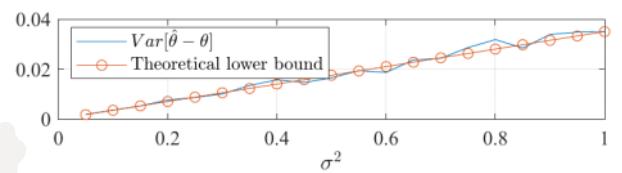
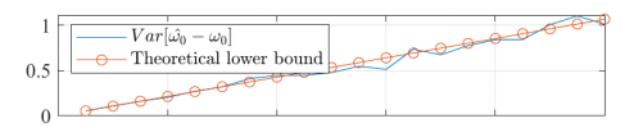


טוקן כיר



$$\text{Var}[\hat{\omega}_0] \geq \frac{12}{SNR \times L(L^2 - 1)} \approx \frac{12}{SNR \times L^3} \quad \left| \begin{pmatrix} rad \\ sample \end{pmatrix} \right. \quad (19.27)$$

$$\text{Var}[\hat{\theta}] \geq \frac{2(2L-1)}{SNR \times L(L+1)} \approx \frac{4}{SNR \times L} \quad [rad^2] \quad (19.28)$$



לעומת זה, מילויים נטויים נרמזים על ידי סימני נטויים בלבד.